Brain and Machine

SOV/29-59-2-4/41

the brain. In modern calculators, the occurrence of accidental signal combinations is avoided. But it is not impossible that such "senseless" combinations play an important part in the activity of thoughts by supplying material for intelligent constructions. If all remembrance cells of the calculator are filled with signals it cannot receive any more informations. A quite specific property of the brain, however, is the ability of comparing the signal combinations stored up at any time, creating economical connections between them and thus making the remembrance susceptible for new informations. Man receives during his life such an abundance of signals that billions of cells would not suffice to keep them in their original form. A generalization of the connections between signals, and the removal of superfluous repetitions, is the aim of the logical function of the brain. A machine capable of removing superfluous repetitions from the stored-up signal combinations without interrupting the connections between them, cannot be designed yet. Besides, the brain can remember its own actions which cannot be achieved by the most perfect machine. The opinion that the principal difference between brain and machine is the fact that the machine works according to a

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Brain and Machine

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program is not correct. Also the brain receives orders and instructions from the body and subordinate parts of the nervous system. The statement that the brain is a machine and that the machine is capable of thinking has to be rejected too. Such statement is an unpardonable depreciation of mental qualities. It justifies imbecility and mental poverty, cruelty and heartlessness toward other people by insisting that there are only reflexes instead of a soul. There are 6 figures.

Card 3/3

SOV/29-59-3-8/23

29(0) AUTHORS:

Gushchev, S., Teplov, L.

TITLE:

How Was It ...? (Kak eto bylo ...?)

PERIODICAL:

Tekhnika molodezhi, 1959, Nr 3, pp 14-17 (USSR)

ABSTRACT:

In this article the authors report on the successful launching of the space rocket on January 2, 1959. Before the rocket was launched, accurate computations of the proper time of launching and the trajectory had to be made with complicated electronic computers. The rocket could not be seen when it took off and only a weak earthquake was noticed. Its flight was watched by locators and in the middle of the screen there was a small bright spot to be seen. After leaving the troposphere the top of the rocket inclined toward the east and the first stage was detached. After the first thirty minutes the rocket had also crossed the ionosphere. Now it was difficult to determine the motion of the rocket from the earth curvature. The computers now did not calculate the trajectory with respect to the earth's surface but with respect to the orbit of the earth. The trajectory of the rocket somewhat declined below this plane. In this moment the next stage was detached and the rocket lost its flashing

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How Was It ...?

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tail. It had attained the parabolic speed of 11.2 km/sec. When the motors stopped, also the last stage weighing 1 1/2 t had reached its trajectory. The top was detached and a rotating ball with aerials and a magnetic feeler on a long rod was hurled out. Each part of the disassembling rocket had the same speed and therefore they flew side by side. Suddenly a seeming deviation from the trajectory was observed on the projection set up according to data calculated by the computers. Yet this was again a paradox of space travelling. Due to the different directions of motion, the projection of the rocket lagged behind the rotation of the earth. After one hour of flight the rocket had already covered a distance of more than 10000 km from the earth. The data of the magnetometer gradually became inaccurate and the magnetic field of the earth became weaker. At a distance of more than 30000 km the rocket had already left the magnetic field of the earth. The instruments in the rocket could only measure cosmic radiation in its original state. Two counters recorded the number of charged particles hitting the rocket, while two photomultipliers analyzed the composition of radiation. After eight hours the rocket had already covered a distance of more than 100000 km. At a distance of 113000 km the quartz clock had

Card 2/3

4

How Was It ...?

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released the fuse of the vaporizer in the right moment, whereby the metallic sodium evaporated in the space as a yellowish cloud several 100 km long. Although radio locators can record the distance from the earth every second, it is yet very difficult to determine the accurate position by radio location. The flash of the sodium cloud that was observed for several minutes permitted a precise determination of the trajectory. Meanwhile the precision instruments in the rocket communicated their observations to the earth. They served the purpose of measuring the degree of density of cosmic gases, and should answer the question whether they are less dense farther from the sun. Further, they recorded the number of corpuscules emitted by the sun. A man, whatever his intelligence, health and reactivity, could never observe and record everything as precisely as these instruments. Therefore, it is unnecessary to risk casualties. At a distance of about 40000km the attractive force of the moon becomes effective. If this line were crossed, the rocket would fall on the moon. The Soviet rocket had another program, however. The instruments communicated the size of the magnetic field and the radioactive intensity of the moon. Communications of the rocket were received still for a long time until it was fully integrated by the solar sphere and became an artificial planet. There are 8 figures.

Card 3/3

TEPLOV, L.

Invisible printing plants. Tekh.mol. 28 no.4:25-26 '60.
(MIRA 13:11)
(Russia--Underground literature) (Printing plants)

Twenty-four hours in space. Time nat. no.10:2 0 '61.

(MIRA 14:10)

(Astronoutics)

	Dreaming at Ap '61.	bout a selftyping	typewriter. Zi	nan. sila 36	no. 4:26-29 (MIRA 14:4)	
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TEPLOV	mhe.	geeing	machine.	Tekh.mol.	30 no.10:5-6 (Perceptrons)	162.	(MIRA 15:12)	
	TIBS	Boorme			(Perceptrons)		•	
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Information, life, death, immortality. Nauka i tekh mladezh 15 no. 11: 3-6 N'63.

ABDULIN, A.; ALEKSEYEV, I.; BANTLE, O.; BOBROV, L.; BOZHANOV, B.;

BOYKO, V.; BONDAREV, K.; BORZOV, V.; VERKHOVSKIY, N.; GUBAREV, V.;

GUSHCHEV, S.; DEBABOV, V.; DIKS, R.; DMITRIYEV, A.; ZHIGAREV, A.;

ZEL'DOVICH, Ya.; ZUBKOV, B.; IRININ, A.; IORDANSKIY, A.;

KITAYGORODSKIY, P.; KLYUYEV, Ye.: KLYACHKO, V.; KOVALEVSKIY, V.;

KNORRE, Ye.; KONSTANTINOVSKIY, M.; LADIN, V.; LITVIN—SEDOY, M.;

MALEVANCHIK, B.; MANICHEV, G.; MEDVEDEV, Yu.; MEL'NIKOV, I.;

MUSLIN, Ye.; NATARIUS YA.; NEYFAKH, A.; NIKOLAYEV, G.; NOVOMEYSKIY, A.;

OL'SHANSKIY, N.; OS'MIN, S.; PODOL'NYY, R.; RAKHMANOV, N.; REPIN. L.;

RESHETOV, Yu.; RYBCHINSKIY, Yu.; SVOREN', R.; SIFOROV, V.; SOKOL'SKIY, A.;

SPITSYN, V.; TEREKHOV, V.; TEPLOV, L.; KHAR'KOVSKIY, A.; CHERNYAYEV, I.;

SHAROL', L.; SHIBANOV, A.; SHIBNEV, V.; SHUYKIN, N.; SHCHUKIN, O.;

EL'SHANSKIY, I.; YUR'YEV, A.; IVANOV, N.; LIVANOV, A.; FEDCHENKO, V.;

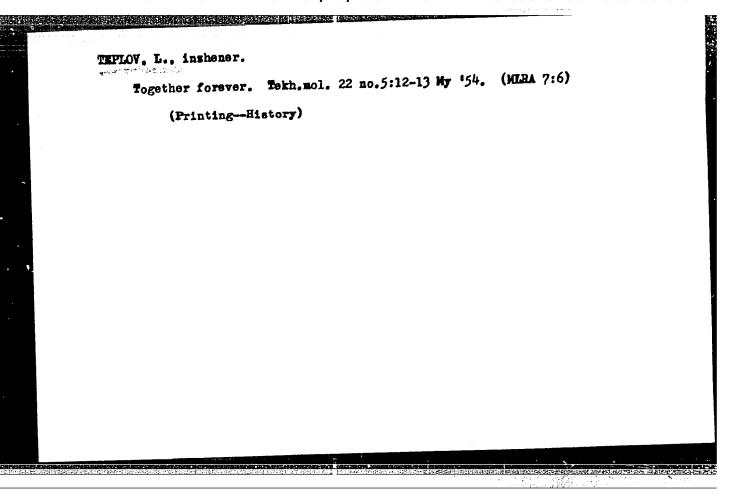
DANIN, D., red.

[Eureka] Evrika. Moskva, Molodaia gvardiia, 1964. 278 p. (MIRA 18:3)

TEPLOV, L.P. (Moscow).

History of Russian printing and publishing; 120 years from the discovery of milticolor planography. Poligr.proizv. no.7:23-24 J1-Ag '53. (MURA 6:9) (Printing-press-History)

APPROVED FOR RELEASE: 07/16/2001 CIA-RDP86-00513R001755310019-9"



TEPLOV, L., inshener.

The machine writes letters. Tekh.mol. 22 no.6:33-35 Je '54. (MLRA 7:6) (Typewriters)

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شند مر دانههمیوسید	Science and technology in countries of people's democracy. Tekh. nol. 23 no.4:26-27 Ap '55. (MLRA 8:6) (Europe, Eastern-Technology) (China-Nuclear physics)	

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TEPLOV.L., inzhener

Electronic photography. Tekh.mol.23 no.7:12-15 J1'55. (MIRA 8:10)

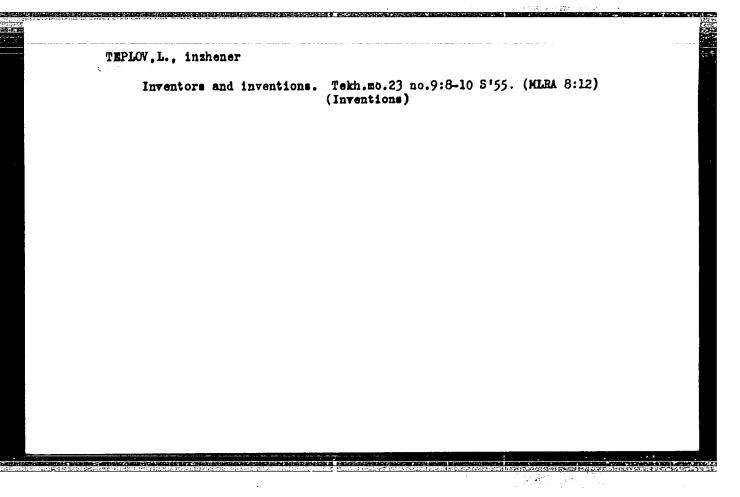
(Printing) (Photography)



Cybernetics. (To be contd.) Nauka i tekh mladezh no.3:25-26 Hr *57.

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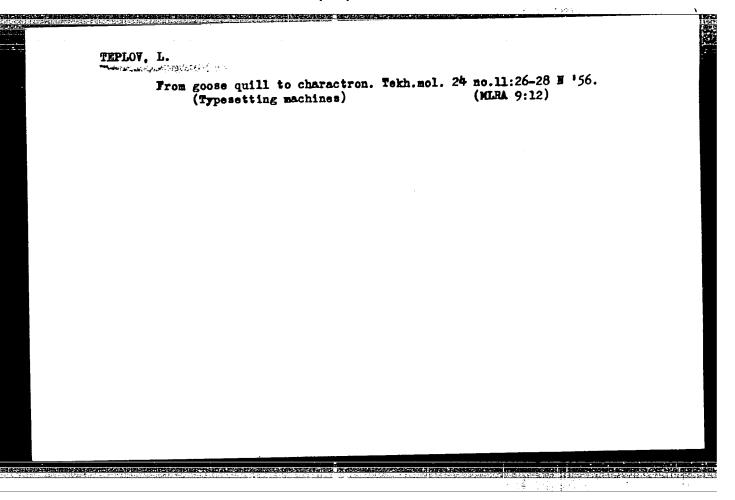
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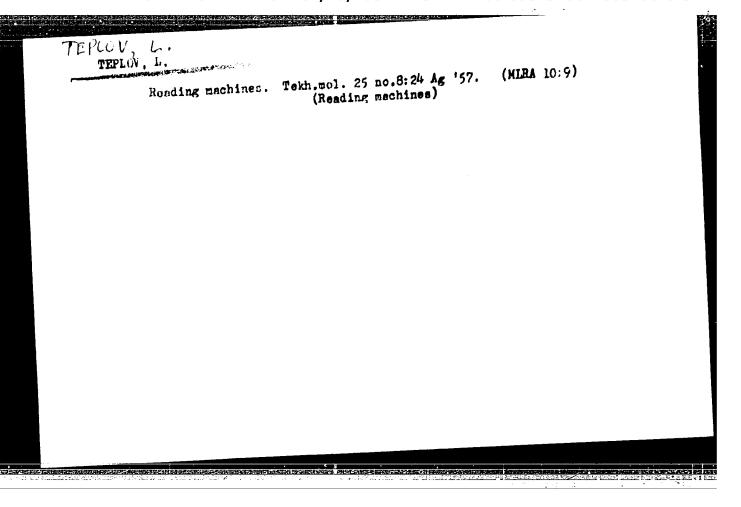
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(Science fiction)

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	(MLRA 10:2)
(Cybernetics)	



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PHASE I BOOK EXPLOITATION

sov/3464

Teplov, Lev Paylovich

Ocherki o kibernetike (Essays On Cybernetics) [Moscow] Moskovskiy rabochiy, 1959. 229 p. 30,000 copies printed.

Ed.: S.Gurov

This book is intended for the layman interested in the science of PURPOSE: cybernetics.

COVERAGE: The book discusses such subjects as the theory of signals, probability theory, biology, physiology of the nervous system, psychology, theory of automatic control, and theory of automatic machines, all of which contribute to the new field of science known as cybernetics. No personalities are mentioned. There are no references.

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AVAILABLE: Library of Congress (Q315.T4)	
	AC/mmp
Card 3/3	AC/gmp 4-26-60

Teplov, Lev Pavlovich

Ocherki o kibernetike. Moskva, Moskovskiy Rabochiy, 1959v. illus.

Bibliographical footnotes.

FOR COMPLETE HOLDINGS CONSULT SHELF LIST

1. Cybernetics. 2. Russia - Cybernetics. 1. Title.

PHASE I BOOK EXPLOITATION

sov/6484

Teplov, Lev Pavlovich

Ocherki o kibernetike (Essays on Cybernetics). 2d ed., rev. [Moscow] Moskovskiy rabochiy, 1963. 413 p. 50,000 copies printed.

Ed.: S. Gurov; Tech. Ed.: Ye. Yakovleva

PURPOSE: This book is intended for the general reader.

COVERAGE: A popular-style description is given of the origin, history, and precent state-of-the-art of cyternatics. Various automatic machines are described, with emphasis on similarities between them and things in nature, and various methods of using, controlling, and accumulating information are outlined. References and recommended reading are given for each chapter and are more or less evenly divided between Soviet and non-Soviet sources.

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Cybernetics, Its Place in Life and Among Sciences

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AL'PEROVICH, Yu.I.; GUTCHIN, I.B.; KAYEYSHEVA, L.S.; TEPLOV, L.P.;

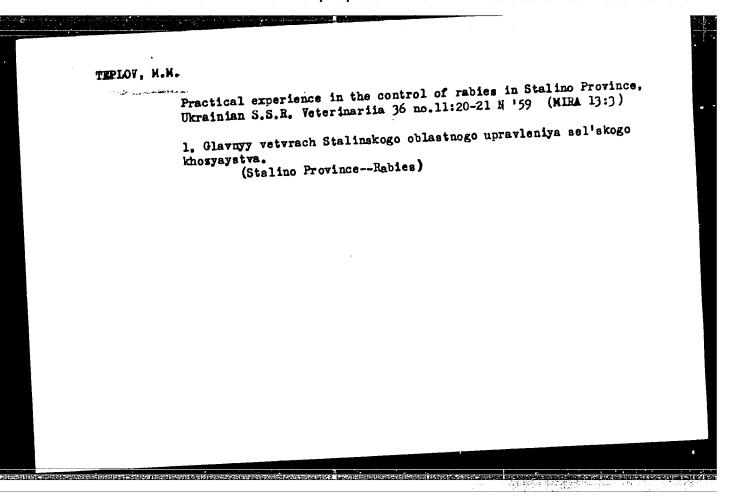
BOCDAMOV, G.C.; DRGITSHEV, Yu.G.; SHENGV, G.V.;

TRET'YAKOV, V.S.; BREYDO, M.I.; YEVSEYEV, L.A.; STEBAKOV,

S.A.; FEDCHENKO, V., red.

[The ABC's of automation; collected articles] Azbuka avtomatiki; sbornik. Moskva, Molodaia gvardiia, 1964. 349 p.

(MIRA 17:7)



TEPHOV, N.L.

AUTHOR: TEPLOY, N.L.

TITLE:

A-U Sci Conf dedicated to "Radio Day", Moscow, 20-25 May 1957. "Basic Correlations in SignalIntegration and Fluctuating Inter-

ference in the Radio Receiver Channel."

PERIODICAL: Radiotekhnik i Elektronika, Vol. 2, No. 9, pp. 1221-1224,

* (USSR). 1957,

For abstract see L.G. Stolyarov.

TEPLOV, N.L.

108-9-1/11

AUTHOR:

On the Evaluation of the Noise Strength of the Radio Reception
Wethoda Rased upon the Masna of Gional and Moise Functions Un the Evaluation of the Noise Strength of the Madlo Receptions
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(K otsenke pomekhoustoychivosti metodov radiopriyema, osnovannykh Teplov, N. L.

TITLE:

na usrednenii funktsiy signala i pomekhi)

PERIODICAL:

Radiotekhnika, 1957, Vol. 12, Nr 9, pp. 3-11 (USSR)

ABSTRACT:

The investigation of the question is here restricted to the general of an averaging linear or animals as the manufacture of an averaging linear or animals. The investigation of the values (discrete and continuous) of the values (discrete and continuous) of the values al case of an averaging linear on principle as the result of a summation of the values (discrete and continuous) of the adding summation of the values (discrete and continuous) of the adding function signal noise with the weight function of function signal noise with the weight function of the adding which (averaging) device, whereby the latter has a constant value for (averaging) device, whereby the latter has a constant value for is here assumed to be equal to 1. The physical supposition at a noise attenuate when the mean of the signal a vield with regard to noise attenuate when the mean of the signal at vield with regard to noise attenuate when the mean of the signal attenuate when the signal attenu 18 nere assumed to be equal to 1. The physical supposition for a yield with regard to noise strength when the mean of the signal and the change noise is taken is practically the complete companies. a yield with regard to noise Strength when the mean of the signal and the chance noise is taken is practically the complete conherence of the signal and an assentially lower scherence of the signal and an assentially lower scherence. and the chance holder is taken is practically the complete co-herence of the signal and an essentially lower coherence of the noise which degree, however, is determined by the selectivity nerence or the signal and an essentially lower concernce of the noise which degree, however, the preponderating of the signal of the total receiver-tract. The preponderating of the noise which is guaranteed by a concrete receiver school of the noise which is guaranteed by a concrete receiver. or the total receiver tract. The prepondersting of the signal over the noise which is guaranteed by a concrete receiver scheme over the noise which is guaranteed by a concrete receiver schemes. over the noise which is guaranteed by a concrete receiver scheme at the input of the recording device different methods of recriterion for the evaluation of the different methods of recriterion for the evaluation of the different methods of recriterion for the evaluation of the different methods of recriterion for the evaluation of the different methods of recriterion for the evaluation of the different methods of recriterion for the evaluation of the different methods of the different metho at the input of the recording device is used as the most gene criterion for the evaluation of the different methods of recriterion for the evaluation

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The state of the s

On the Evaluation of the Noise Strength of the Radio 108-9-1/11 Reception Methods Based upon the Means of Signal- and Noise Functions.

ception from the point of view of reliability. Pollowing methods are discussed:

- 1) Method of repeated repetition realizes the idea of accumulation in the most simple way. The single measurements or the readings of every mixture of signal and noise at all which repeat with a period T are added and the mean is teaken. It is shown that the yield in the case of a reponderance of the signal over the noise is equal to the number of the repetitions n.

 2) The integrating of the signal and the noise can from the physical point of view be described as a continuous summation the effective range of the signal. A formula is derivated which regard to the noise strength in the case of integrating is directnoise correlation interval which is determined by the breadth of Discrete teleform the receiver tract.
- 3) Discrete taking of the mean of the signal and the chance noise. The investigation is restricted here to the case of a constant signal level.

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On the Evaluation of the Noise Strength of the Radio 108-9-1/11 Reception Methods Based upon the Means of Signal- and Noise Functions.

4) Method of accumulation. This is considered as a pairing of the repeated repetition with the integrating of the signal and the noise within the effective range of the single-signal. It is shown that in the case of broad bands the efficiency of the method of taking the mean in a concrete apparatus is determined only by the total signal effective time T and does not depend practically upon the fact wether the signal is transmitted discretely or continuously. There are 8 figures and 2 Slavic references.

SUBMITTED: October 31, 1956 (initially) and January 17, 1957 (after revision)

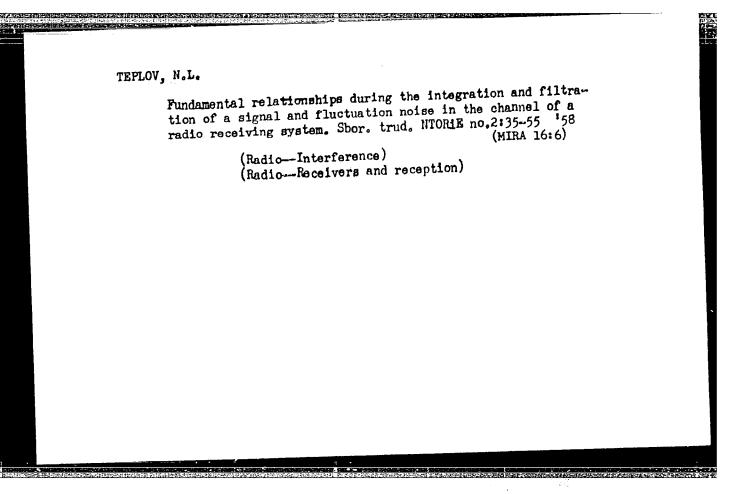
AVAILABLE: Library of Congress

Card 3/3

TEPLOV, N. L.

N. L. Teplov, "Certain questions of the theory and computation of the interference immunity of impulse radio reception." Scientific Session Devoted to "Radio Day", May 1958, Trudrezervizdat, Moscow, 9 Sep 58.

Questions of the maximum approximation of interference-immunity of the radio reception of impulse signals to the potential interference immunity are analyzed as is also a general method to analyze and compute the interference immunity of impulse radio reception circuits.



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507/106-59-1-4/12

AUTHOR: TITLE:

Teplov, N.L.

The Maximum Noise Stability of Radio Reception of Signals with Amplitude, Frequency and Phase Keying (Maksimal naya

pomekhoustoychivost radiopriyema signalov s amplitudnoy,

chastotnoy i fazovoy manipulyatsiyey)

PERIODICAL: Elektrosvyaz 1959, Nr 1, pp 28-37 (USSR)

ABSTRACT: Fig 1 shows a typical AM receiver, while (1) are equations for the mark and space versions of the signal. The

receiver is working correctly when its output is greater than a certain threshold value during the signal period and less during the space period. Errors occur when the situation is reversed. The probability density of the envelope of the sum of sinusoidal signal of fluctuation noise is given by (2). The mean square value of the noise in the effective receiver bandwidth is (3) and the

probability density of the noise in the absence of signal is given by (4). For equal a priori probabilities of mark and space the probability of error during reception is given by (5). Combining (3) and (4) into this last

equation gives (6). The error probability is reduced by increasing the signal/noise ratio at the input to the

detector and for a given signal/noise ratio the

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The Maximum Noise Stability of Radio Reception of Signals with Amplitude, Frequency and Phase Keying

> probability of error depends on the threshold value. A study of the minima of the function in (6) reveals an optimum value for the threshold (8) for a sufficiently This gives the commonly used large signal/noise ratio. design value in (10) (limiting level equal to one half expected signal level). Eq (8) is plotted in Fig 2. The minimum probability of error when receiving AM signals with an optimum threshold value is given by (11). It will be seen that this probability is uniquely determined by the signal/noise ratio at the input to the detector. The action of frequency shift keying is defined in (12) and a block diagram of a suitable receiving system is in Errors occur in the system when the value of the Fig 3. noise envelope coming out of one separating filter through which there is no signal exceeds the value of the total envelope for signal-plus-noise coming out of the filter in which there is a signal. The analysis proceeds as before and the probability of error is given by (16).

Card 2/5 Frequency shift keying is defined in (17) and a block diagram of a suitable receiver is in Fig 4.

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The Maximum Noise Stability of Radio Reception of Signals with

Amplitude, Frequency and Phase Keying

fluctuation noise at the input to the separating circuit takes the form of a sinusoidal oscillation whose envelope and phase are slowly varying functions of time. The corresponding probability of error is given by (24). It is stated by the author that the formulae (11), (16) and (24) were obtained by him in 1954 in fulfilment of his dissertation work. The penultimate section is devoted to a calculation of the maximum noise stability of radio reception of signals with these various forms of keying. Eq (27) is the expression of the square of the signal/ noise ratio. The optimum bandwidth of the filter for passing a rectangular pulse has been found by Siforov (Ref 4) as (30), while the optimum bandwidth for a single tuned circuit is given by Teplov (Ref 5) as (31). The corresponding maximum signal/noise ratios are (32) and (33). The formulae are repeated in (34) and (35) in terms of Q2, which is the ratio of the energy in an elementary signal to the noise density at the input. useful concept is introduced which is the limiting Card 3/5 signal/noise ratio in reception obtained by integrating

The Maximum Noise Stability of Radio Reception of Signals with Amplitude, Frequency and Phase Keying

for the duration of the signal. It is shown in the simple analysis in the top half of page 34 that the simple analysis in the top half of page 34 that the limiting signal/noise ratio is in fact equal to Q² (Eq 42). limiting signal/noise ratio is in fact equal to Q² (Eq 42). limiting signal/noise ratio is in fact equal to Q² (Eq 42). applying this last expression to the three particular applying this last expression to the problem by (43), (44) cases the noise stability for each is given by (43), (44) and (45) for AM, FM and PM respectively. The last section gives formal solutions to the problem of radio reception in such a way as to realise the potential noise stability as defined by Kotel'nikov (Ref 6). Block diagrams for PM, FM and AM are Figs 5, 6 and 7 respectively. Corresponding expressions for potential noise stability are (46), (52) and (56). Graph of Fig 8 shows the dependence of the probability of error on Q when receiving the three kinds of signal with synchronous

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The Maximum Noise Stability of Radio Reception of Signals with Amplitude, Frequency and Phase Keying

and amplitude detectors. By using coherent detection the maximum noise stability equals the so called

potential value. There are 8 figures and 6 Soviet references.

SUBMITTED: June 30, 1958

Card 5/5

20475 S/106/61/000/004/002/004 A055/A133

9,3275 (also 1031, 1067)

6.4400 AUTHOR:

Teplov, N. L.

TITLE:

The noiseproof feature of integrated reception of signals in the case of fluctuation noises and of sinusoidal interferences

PERIODICAL: Elektrosvyaz', no. 4, 1961, 9-18

TEXT: Integrated reception is compared, as a rule, to the usual narrow-band reception. Such a comparison is natural enough, since integration (averaging), considered as a physical process, is equivalent to filtration (smoothing). By narrow-band reception is usually understood the reception with the most advantageous pass-bands corresponding to the optimum noiseproof with the most advantageous pass-bands corresponding to the optimum frequency bands, for reception. As to fluctuation noises, the most advantageous pass-bands, for reception of single pulses, are the so-called optimum frequency-bands (V. I. Siforov - Ref. 1: "Influence of Noises on the Reception of Pulse-Signals", Radiotekhnika, 1946, no. 1). The author of the present article begins with some general considerations regarding the main features of the integrated reception, as compared to the usual narrow-band reception. He points out that: 1) the noises exert an effect upon the integrating circuit only

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The noiseproof feature of integrated...

within the duration of the signal, which is not the case in the usual narrow-band reception, where filters are "static" or permanently switched on; 2) in integrated reception, the superposition of adjacent pulses, due to residual oscillations, is eliminated; 3) the salient difference between the two reception methods is the fact that, in the integrated reception, the frequency selectivity is determined, not by "static", but by "dynamic" resonance characteristics and pass-bands of the integrating circuits. After these general considerations, the author proceeds to a comparative theoretical analysis of the noiseproof feature in both reception methods. He first examines the case of fluctuation noises. The principal formula used by him is the formula giving the ratio of the signal-power to the noise-power at the output of the integrating circuit:

of the integrating circuit: $h_{*}^{2} = \frac{a_{7}^{2}}{b_{7}^{2}} = \frac{a^{2}\tau}{2v^{2}}$ (14)

where $a_{\mathcal{T}}$ is the amplitude of the signal at the output of the circuit, \mathcal{T} is the duration of the signal, $v_{\mathcal{T}}^2$ is the intensity of the noise in the 1-cycle band (specific intensity), and $b_{\mathcal{T}}$ expresses quantitatively the amplitude of

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20475 \$/106/61/000/004/002/004

The noiseproof feature of integrated ...

the noise. For the usual narrow-band reception, the corresponding formula iв

 $h_{opt}^2 = 0.82 \frac{a^2 \tau}{2\sqrt{2}}$ (16)

This formula (16) is valid for the reception of single pulses and for any filter with the optimum frequency-band. For the extreme cases of an ideal filter and of a single oscillating circuit, formula (16) becomes $h_{\text{opt.id.f.}}^{2} = 0.825 \frac{\text{a}^{2} c}{2\sqrt{2}}$ (21)

and

 $h_{\text{opt.single osc.}}^2 = 0.815 \frac{a^2 c}{2^{\sqrt{2}}}$ (19)

respectively.

Taking the ratio of (14) to (16), the author finds: $\left(\frac{h_*}{h_{opt}}\right)_{fluct}^2 = 1.22$ (22)

For the reception of single, widely spaced pulses, the integrating reception method is, therefore, almost equivalent to the usual narrow-band method with optimum frequency-bands. But the advantage offered by the integrating method

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The noiseproof feature of integrated ...

20475

proves more substantial in the case of an uninterrupted sequence of signalpulses (for instance, in radiotelegraphy). In this case, a short mathematical

 $h_{\text{opt.}}^2 \approx 0.5 \frac{a^2 \tilde{\chi}}{2 v_0^2}$ (26)

The ratio of (14) to (26) is: $(\frac{h*}{hopt.})^2$ fluct. ≈ 2 As to the ratio of signal-power to noise-power, the integrated reception of an uninterrupted train of pulses ensures thus a gain practically equal to two, in comparison with the usual narrow-band reception. This gain proves still greater in the case of signals subject to fading. In the last part of the article, the author examines the noiseproof feature of both reception methods in the case of sinusoidal interferences. Here also, he deduces formulae giving h_{*}^{2} , h_{opt}^{2} , and h_{opt}^{2} , and finds that: $\left(\frac{h_{*}}{h_{opt}}\right)_{sin}^{2} = 1.22 \tag{39}$

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20475 S/106/61/000/004/002/004 A055/A133

(40)

The noiseproof feature of integrated ...

and
$$(\frac{h_\#}{h_{\text{opt}}})_{\text{sin}}^2 \approx 2$$
 There are 6 figures and 5 Soviet-bloc references.

SUBMITTED: April 12, 1960

Card 5/5

80162

S/108/60/015/04/04/007 B014/B014

6,9000

AUTHOR:

Teplov. N. L., Member of the Society

TITLE:

The Maximum Noiseproof Feature in Systems With Discrete Signals 8

PERIODICAL: Radiotekhnika, 1960, Vol. 15, No. 4, pp. 27 - 35

TEXT: In the article under review, the author studies the maximum noiseproof feature of systems having discrete signals in the case of coherent and incoherent reception. Formula (2) is written down for the probability of regular signal detection. The fluctuating noise is expressed by formula (3) which is split up into its components according to (4). Then, the author gives formula (8) for the correlation function between the amplitudes of formula (4). Formula (8) is used to determine the energy spectrum of the fluctuating noise. The relation between the signal expressed by (1) and the fluctuating noise may be represented by the vector diagram shown in Fig. 2. Next, the author carries out a general estimation of the noiseproof features of systems with discrete signals in the case of coherent and incoherent reception. In both cases he proceeds from the probability that the amplitude of the noise in branches without signals is not larger than the amplitude of signal plus noise in branches with signals. This is described by

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80162

The Maximum Noiseproof Feature in Systems With Discrete Signals

S/108/60/015/04/04/007 B014/B014

formulas (13) and (20). Subsequently, he develops formulas (18) and (24) for the probability of regular signal detection for the two cases under consideration. The two formulas assume their definite form with formulas (27) and (28). When investigating the maximum noiseproof feature the author first studies the maximum surpassing of the signal over the noise. Integrals (43) and (44) are given for the determination of the maximum noiseproof feature. Formula (47) is used to determine the relative error arising in the estimation of the noiseproof feature. In conclusion, the author compares the noiseproof features in systems with discrete signals in the case of coherent and incoherent reception. Table 2 indicates that the greatest difference in the noiseproof features of the two modes of reception is obtained in the range of the highest values of error probability. As compared to incoherent reception, the greatest gain of signal power in the case of coherent reception is equal to 2. Hence, when signals with a given error probability are received, the signal voltage at the input of an incoherent receiver must be higher than that at the input of a coherent receiver. The respective factor between the two input voltages varies from 1 to 12. In conclusion, the author gives the formulas for the maximum noiseproof feature and those for the probability of detecting signals for the case in which fluctuating noise occurs only in one part of the branches of this system: formulas (56), (58), and (60).

Card 2/3

"APPROVED FOR RELEASE: 07/16/2001

CIA-RDP86-00513R001755310019-9

The Maximum Noiseproof Feature in Systems With Discrete Signals

80162 S/108/60/015/04/04/007 B014/B014

This article was read at the All-Union Anniversary Session of the NTORIE imeni A. S. Popov in June 1959. There are 5 figures, 2 tables, and 3 Soviet references.

SUBMITTED: March 6, 1958 (initially) and July 24, 1959 (after revision)

4

Card 3/3

31197 S/106/61/000/012/001/010 A055/A127

6.9400

AUTHOR:

Teplov, N. L.

TITLE:

Noise immunity in integrated reception of signals in the presence of impulse or transient sinusoidal interferences

PERIODICAL: Elektrosvyaz', no. 12, 1961, 3 - 12

TEXT: This article is a theoretical comparison of the noise immunity of the integrated reception method and the usual narrow-band reception method in the case of: 1) impulse interferences, 2) transient (short-term) sinusoidal interferences. 1) Impulse interferences. a) Narrow-band reception. - The impulse interferences at the filter output are given by the expression:

 $C(t) = c(t)\cos(\omega_0 t + \gamma_0). \tag{1}$

 ω_0 is here the angular frequency of the tuned filter; φ_0 is determined by the moment when interferences appear; c(t) is the amplitude (envelope) of the oscillations, whose time-variation determines the shape of the impulse interference at the filter output. Starting from this formula, the author finds the ratio between the maximum amplitudes of the signal and the impulse interference single pulses at

Card 1/6

Noise immunity in integrated reception of ...

311)7 3/105/61/000/012/001/010 A055/A127

the output of the narrow-band filter. In the case of a single oscillating circuit, this ratio is:

$$\left(\frac{a_{\text{outp max}}}{c_{\text{outp max}}}\right)_{\text{single circ}} = \frac{a(1-e^{-2\Delta f}eff^{T})}{2 S_{0}\Delta f_{eff}},$$
(5)

where S_0 is the modulus of the spectral density of the impulse interference at the filter input; $\Delta f_{\rm eff}$ is the effective frequency-band of the filter; τ is the duration of the signal pulse. At $\Delta f_{\rm eff}$ opt. = $\frac{0.65}{\tau}$ (for single signal pulses):

$$\frac{\left(\frac{\text{a outp max}}{\text{coutp max}}\right)}{\text{coutp max opt. single circ.}} = \frac{\frac{a}{3_0} \left(\frac{(1-e^{-1.3})}{2 \cdot 0.65} = 0.56 \frac{\text{a}}{3_0}\right)}{2 \cdot 0.65} = 0.56 \frac{\text{a}}{3_0}. \tag{7}$$

At Δf^{*}_{eff} opt = $\frac{1.1}{C}$ (for reception of an uninterrupted train of pulses):

$$\frac{\left(\frac{\text{a_outp max}}{\text{c_outp max}}\right)_{\text{opt'single circ.}} = 0.36 \frac{\text{aV}}{\text{S}_{\text{o}}}.$$
(8)

Analogous formulae are then derived for ideal band-filters. In the general and average case, it can be assumed that:

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31197 3/106/61/000/012/001/010 A055/A127

Noise immunity in integrated reception of ...

$$\left(\frac{a_{\text{outp max}}}{c_{\text{outp max}}}\right)_{\text{opt}} \approx 0.5 \frac{aT}{S_0}.$$
(12)

$$\left(\frac{a_{\text{outp max}}}{c_{\text{outp max}}}\right)_{\text{opt}} \approx 0.5 \frac{aT}{S_0}.$$

$$\left(\frac{a_{\text{outp max}}}{c_{\text{outp max}}}\right)_{\text{opt}} \approx 0.3 \frac{aT}{S_0}.$$
(12)

b) Integrated reception: Δf_{eff} being here the effective static band of the integrator filter, the author shows that the maximum ratio between the signal and the impulse interference is determined by the formula:

$$\left(\frac{a}{c}\right)_{integr.} = \frac{aT}{S_0}$$
 (14)

Comparing (14) with (12) and (13), the author obtains: for the reception of single pulses:

$$\frac{\left(\frac{a}{c}\right)_{\text{integr}}}{\left(\frac{a}{c}\right)_{\text{opt}}} = 2; \tag{20}$$

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CIA-RDP86-00513R001755310019-9" **APPROVED FOR RELEASE: 07/16/2001**

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Noise immunity in integrated reception of ...

S/106/61/000/012/001/010 A055/A127

for the reception of an uninterrupted train of pulses:

$$\frac{\left(\frac{a}{c}\right)_{\text{integr}}}{\left(\frac{a}{c}\right)_{\text{opt}}} \approx 3. \tag{21}$$

This comparison shows the advantage of using the integrated reception. 2) Transient (short-term) sinusoidal interferences. a) Integrated reception: The average statistical amplitude of the interference is assumed to be the same and equal to einterf. The ratio between the signal and the transient sinusoidal interference is given by:

 $h_{*}^{2} = \frac{a_{\text{integr}}^{2}}{\frac{e^{2}}{\text{interf}}} = \frac{a^{2}\tau}{e_{\text{interf}}^{12}} \frac{\tau}{\tau_{\text{interf}}} = \frac{a^{2}\tau}{e_{\text{interf}}^{2}} \frac{\tau}{\tau_{\text{interf}}}, \quad (23)$

where \mathcal{T}_{interf} is the duration of each interference and ξ_0^2 is the square of the effective value of the interference, assumed to be the same at my frequency. b) Narrow-band reception: The signal-to-interference ratio at the output of the narrow-band resonance circuit is determined by:

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S/106/61/000/012/001/010 A055/A127

Noise immunity in integrated reception of ...

$$\left(\frac{a^{2}}{e^{2}_{interf}}\right)_{outp} = \frac{a^{2}(1-e^{-2\Delta f_{eff}T})^{2}}{e^{2}_{interf} \Delta f_{eff}(1-e^{-4\Delta f_{eff}T_{interf}})}$$
(29)

where Δf_{eff} is the effective frequency-band of the circuit. If $\Delta f_{eff,opt} = \frac{0.65}{7}$,

$$\frac{\left(\frac{a_{\text{outp}}^{2}}{e_{\text{interf}}}\right)_{\text{opt}} = \frac{a^{2}(1-e^{-1.3})^{2}}{e_{\text{interf}}^{1/2}\left(1-e^{-2.6T_{\text{interf}}/T}\right)} = \frac{a^{2}\pi}{e_{\text{interf}}^{1/2}} = \frac{0.82}{e_{\text{interf}}^{1/2}}$$
(31)

Comparing formulae (23) and (31), the author writes:

$$A = \frac{(23)}{(31)} = 1.22(1 - e^{-2.6p})\frac{1}{p},$$
 (33)

where $p = T_{interf}/T$. Formula (33) permits to rate the advantages of the integrated reception. After examining the effect of the irregular variations of the signal pulse amplitude, the author draws the following conclusions: The noise immunity of the integrated reception, as compared to the narrow-band reception, is the greatest in the case of impulse interferences and transient sinusoidal interferences.

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Noise immunity in integrated reception of ...

ces. In the reception of single signal pulses, the immunity of the integrated reception as regards fluctuation and "undamped" sinusoidal interferences is but little different from that of the usual narrow-band reception with optimum frequency-bands. The integrated reception noise immunity for each kind of interferences is determined by the magnitude of the signal-to-interference ratio; these magnitudes, such as computed by the author, are listed in two tables. There are 2 figures, 2 tables and 3 Soviet-bloc references. The names of Soviet scientists mentioned in the article are Kotel'nikov and Conorovskiy.

SUBMITTED: April 18th, 1960

Card 6/6

TEPLOV, N.L.

Determination of the functions of an ideal receiver. Addiotokinika 16 no.3:31-39 %r '61. (M. A 14-2)

1. Doystvitel'nyy chlen Mancinc-tekiniekoskogo obshekestva redictekiniki i elektrosvyazi im. A.S.Popova. (Radic-Recievers and reception) (Information theory)

39166 \$/106/62/000/008/001/009 A055/A101

9.3280

AUTHORS: Teplov, N.L., Shmatchenko, V.F.

TITLE: Analysis of the integrator of rectangular radio pulses

PERIODICAL: Elektrosvyaz', no. 8, 1962, 3 - 12

TEXT: The authors determine the parameters of the ideal integrator (with linear integration) and of the integrator with a single oscillating circuit. Ideal integrator: The equation of the integrator resonance characteristic is:

$$y \left(\Delta f\right)_{\tau} = \frac{a \left(\Delta f\right)_{\tau}}{a \left(0\right)_{\tau}} = \frac{\sin \pi \Delta f \tau}{\pi \Delta f \tau} , \qquad (9)$$

where $\Delta f = \frac{\Delta \omega}{2\pi}$ and $\Delta \omega = \omega - \omega_0$, t is the duration of the integration. The effective frequency-band of the integrator is

$$\Delta f_{\text{eff }\tau} = \int_{-\infty}^{\infty} y^2 (\Delta f)_{\tau} d\Delta f = 2 \int_{0}^{\infty} \frac{\sin^2 \pi \Delta f \tau}{(\pi \Delta f \tau)^2} d\Delta f = \frac{1}{\tau}.$$
 (12)

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S/106/62/000/008/001/009 A055/A101

Analysis of the integrator of rectangular

Assuming that the signal is A (t) = $a_0 \sin \omega_0$ t at $0 \le t \le \tau$, (13) (τ being the duration of the signal), the excess of the signal over the interference at the output of the integrator will be

$$h_{\text{id integr}}^{2} = \frac{\frac{a_{0}^{2}}{2}}{v_{0}^{2} \Delta f_{\text{eff}} \tau} = \frac{a_{0}^{2} \tau}{2 v_{0}^{2}}, \qquad (14)$$

where v_0^2 is the specific intensity of the interferences (in the 1-cycle-band). Integrator with a single oscillating circuit: The authors obtain the dynamic resonance characteristic of the intergrating circuit:

$$y (\Delta f)_{\tau} = \frac{a (\tau, \Delta f)}{a (\tau, \Delta f = 0)} = \frac{\sqrt{1 - 2 e^{-2\gamma} \cos 2\pi x + e^{-4\gamma}}}{\sqrt{1 + (\pi \frac{x}{\gamma})^2 (1 - e^{-2\gamma})}},$$
 (21)

where $\gamma = \tau \Delta f_{eff}$; $x = \Delta f \tau$. At $\tau \rightarrow \infty$,

$$\lim_{\tau \to \infty} y \left(\Delta f \right)_{\tau} = y \left(\Delta f \right) = \frac{1}{\sqrt{1 + \left(\pi \frac{x}{y} \right)^2}}, \tag{22}$$

Card 2/4

Analysis of the integrator of rectangular

S/106/62/000/008/001/009 A055/A101

which is the equation of the static resonance characteristic. On the basis of (21), the authors calculate the ordinates corresponding to the minima and maxima of the dynamic characteristic. They next give the formula for the effective dynamic frequency-band of the integrator:

$$\Delta f_{\text{eff }\tau} = \int_{-\infty}^{\infty} y^2 (\Delta f)_{\tau} d\Delta f = \int_{-\infty}^{\infty} \frac{1 - 2 e^{-2\gamma} \cos 2\pi \Delta f \tau + e^{-4\gamma}}{\left[1 + (\pi \frac{\Delta f \tau}{\gamma})^2\right] (1 - e^{-2\gamma})^2} d\Delta f. \quad (28)$$

This formula shows that, at an unlimited narrowing of the static band, the dynamic frequency-band of the integrating circuit tends towards the effective frequency-band of the ideal integrator. The excess of the signal over the interferences at the output of the integrating circuit is:

$$h_{\text{integr}}^{2} = \frac{a_{0}^{2}}{2 v_{0}^{2} \Delta f_{\text{eff }\tau}} = \frac{a_{0}^{2} \tau}{2 v_{0}^{2}} \frac{1}{\gamma} \frac{1 - e^{-2\gamma}}{1 + e^{-2\gamma}} = Q^{2} \frac{1}{\gamma} \frac{1 - e^{-2\gamma}}{1 + e^{-2\gamma}}, \quad (32)$$

where $Q^2 = h^2$ is the limit-value of the excess, and A_0 is the input signal ampli-

Card 3/4

S/106/62/000/008/001/009 A055/A101

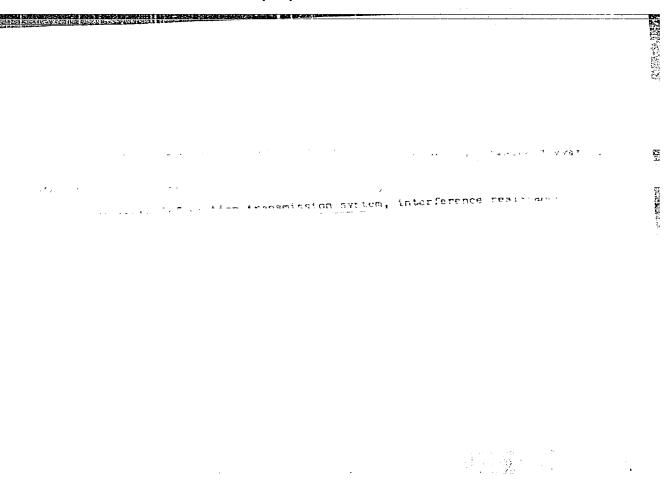
Analysis of the integrator of rectangular

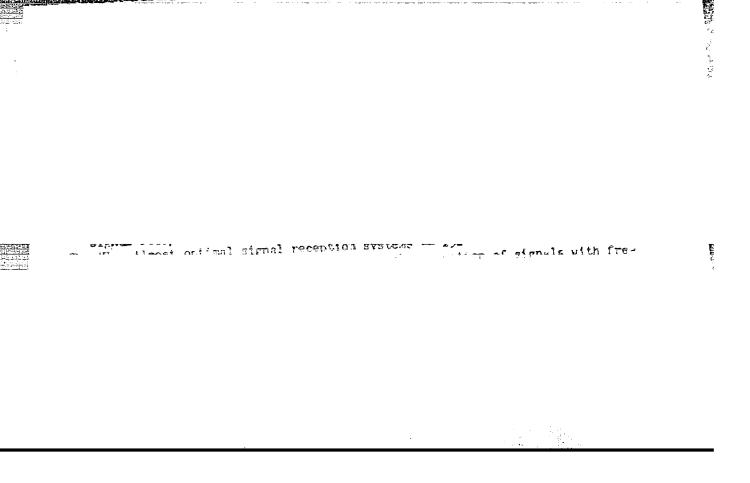
damping of the oscillations is k=e , where $\alpha_{\Sigma}=\alpha_1+\alpha_2$ (α_1 being the attenuation of the circuit in the integration period, and α_2 the additional attenuation for the damping of the oscillations), and $\Delta \tau$ damp is the damping attenuation for the damping of the effective frequency-band of the circuit in time. Designating by $\Delta f_{\rm eff} \Sigma$ the effective frequency-band or (replacing damping operation, the authors write: $\ln k = 2\Delta f_{\rm eff} \Sigma$ $\Delta \tau$ damp, or (replacing damping operation, the authors write: $\ln k = 2\Delta f_{\rm eff} \Sigma$

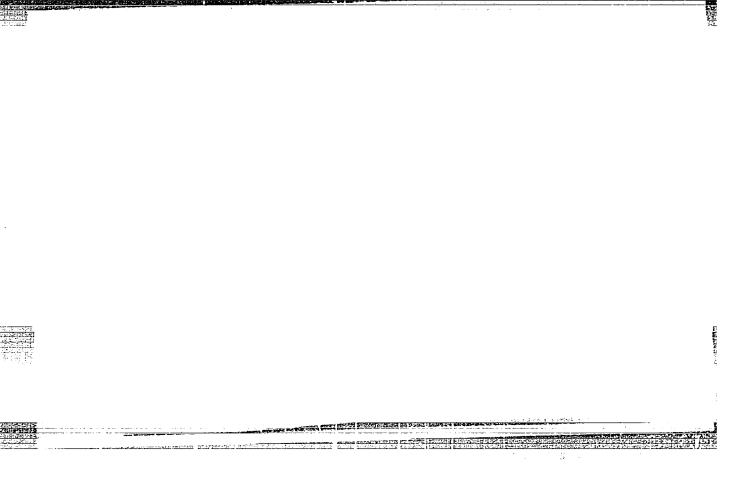
 $\Delta f_{\rm eff} \simeq \Delta^{\rm t}_{\rm damp}$ by $\Delta \gamma_{\rm Z}$): $\Delta \gamma_{\rm E} = \frac{11.5}{2}$. A graph illustrates the dependence (36). The Soviet personalities mentioned in the article are: V.A. Kotel'nikov, I.S. Gonorovskiy. There are 11 figures.

SURWITTED: January 13, 1962

Card 4/4







"APPROVED FOR RELEASE: 07/16/2001 CIA

CIA-RDP86-00513R001755310019-9

L 7646-66 EWI(1)/EWA(h)

ACC NR: AP5024996

SOURCE CODE: UR/0286/65/000/016/0059/0060

AUTHOR: Teploy, N. L.

19

ORG: none

TITLE: Kinematic filter. Class 21, No. 173857

SOURCE: Byulleten' izobreteniy i tovarnykh znakov, no. 16, 1965, 59-60

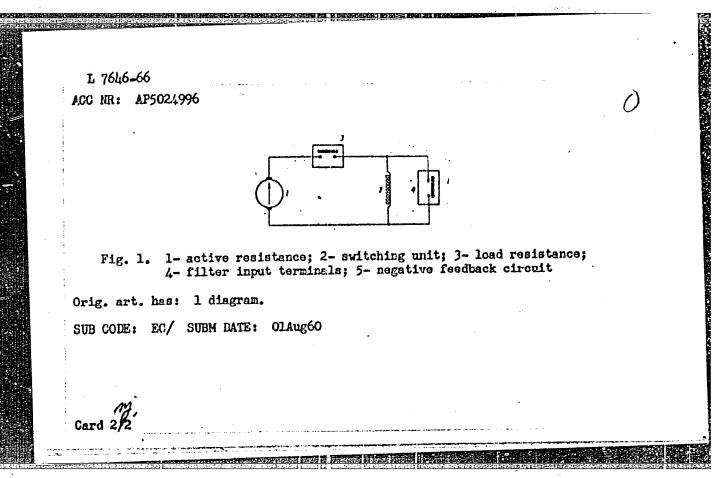
TOPIC TAGS: filter circuit, resonator

ABSTRACT: This Author Certificate presents a kinematic filter made of a quartz resonator connected to the load and provided with an active resistance for suppressing residual escillations. To increase the effectiveness of residual escillation suppression, the active resistance is connected in series with a switching unit and in parallel with the load resistance (see Fig. 1). A negative feedback circuit is connected between the cutput and input of the filter.

Card 1/2

UDC: 621.372.543.2

2



ACC NRI AT6022363

SOURCE CODE: UR/0000/66/000/000/0012/0016

AUTHOR: Teploy, N. L.

ORG: none

TITLE: Potential noise rejection ability and methods of realization of optimal codes

SOURCE: Vsesoyuznaya nauchnaya sessiya, posvyashchennaya Dnyu radio. 22d, 1966. Sektsiya teorii informatsii. Doklady. Moscow, 1966, 12-16

TOPIC TAGS: signal noise separation, signal coding

ABSTRACT: Examination of known theoretical formulas for errors in code-transmission systems shows that: (1) The noise-rejection ability of an optimal (equidistant) binary code is determined only by its base m and is independent of the number n of digits used; (2) Within acceptable error-probability values $(10^{-5}--10^{-6})$, the orthogonal code having $m_{\rm max}$ ensures a probability of information-transmission error by three orders of magnitude lower than that of the optimal binary code. Methods of constructing orthogonal codes are briefly discussed in general terms. Orig. art. has: 1 figure and 12 formulas.

SUB CODE: 17, 09 / SUBM DATE: 28Apr66 / ORIG REF: 004 / OTH REF: 000

Card 1/1

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Teodoravich, V.P.; Teplov, N.S.

TITUM: Effect of gas-saturated hyper on dis acrongth and dectility characterisation of titanium alloys

SURCE: AN SSSR. Institut metallurgit. III allyego uplays, no. 10, 1963.
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TEPIOV, N.V.

Studying the specificity of antigons from Bunostomia on quinea pigs and sheep. Veterinariia 41 no.4:51-52 F **165. (1984-18:3)

1. Biologo-pochvennyy institut Dal'hevostochnego flijela Sibirakego otdeleniya Ali SOSR.

S/191/61/000/008/002/006 B110/B201

AUTHORS:

15.8130

Kamenskiy, I. V., Itinskiy, V. I., Teplov, N. Ye.

'Andrianov, B. V.

TITLE:

Synthesis and study of monomeric and polymeric reaction

products of acetophenone with furfurole

PERIODICAL:

Plasticheskiye massy, no. 8, 1961, 12 - 15

TEXT: Reaction products of acetophenone with furfurole are as follows:

$$\begin{array}{c} O & CHO + CH_3 - CO - C_0H_3 \\ \hline & O & CH = CH - CO - C_0H_3 \\ \hline & -H_2O \\ \hline & O & CHO + 2CH_3 - CO - C_0H_3 \\ \hline \end{array}$$

Card 1/8

O CH CH - CO - CH (II)

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Synthesis and study of monomeric...

(II) is obtained with considerable excess of acetophenone only. (I) is prepared by condensation of equimolecular amounts of furfurole and acetophenone by means of sodium ethylate in alcohol in a yield of 60 - 80 %. In consideration of the fact that the production of resins by means of benzene sulfonic acid catalysts and resulting resin products had been hitherto insufficiently described, their description was the aim of the present work. The authors used (1) furfurole, (2) acetophenone. The polymers were obtained (I) directly from the reaction mass without separation from monofurfurylidene acetophenone (MFAP), (II) by way of resinification of MFAP. The product produced by Harvey's method (Ref. 8: USA Patent 2,461,510 (1949)) loses fluidity on the passage to the B stage. Hardening takes place at 250°C during 30 minutes with the separation of 50% of volatile parts. The authors washed the reaction mass with cold water, dried it at 100°C and 15 mm Hg during 3 hr, thus obtaining a brown oily liquid. A vacuum distillation yielded: 14% furfurole, 16% acetophenone, 60% MFAP, 10% resin. After 3.5 hr of heating at 250°C a fusible black resin (dropping point 65°C) was obtained. On addition of 5% benzene sulfonic acid (50% acetone solution) the resin is hardened during 19 minutes at 250°C under separation of 40% of volatile parts and Card 2/8

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formation of foaming products. Table 1 shows that in MFAP production under optimum, equimolecular conditions, a temperature drop (experiments 1 - 5) reduces the resin formation and at the same time retards the MFAP formation. An increase of the catalyst amount (experiments 4, c. 7.9-12) and a concentration increase of its aqueous solution (experiments 10 - 11), however, speed it up. At room temperature (experiments 8 - 12), MFAP is obtained without resin. 20 g KOH in 20 g H₂O were added by drops to 96 g furfurole and 120 g acetophenone within 20 2 30 minutes, neutralized with 0.5 N HCl, washed with H20 until Cl ions were removed completely, and dried in vacuum. MFAP is bright-yeliow, fine-crystalline with the melting point 41.8°C, and 89 % of the theoretical yield, soluble in all organic solvents (to 12 % in petroleum ether). Its specific gravity was 1.1120, the boiling The molecular weight. point 186°C at 11 mm Hg, 181°C at 9 am Hg cryoscopically determined in dioxan was 196.8, the oxime number was obd. since benzylidene acetophenone compounds and two hydroxyl amine molecules Resinification took place (I) thermally, (II) in the presence of a catalyst According to Table 3, resinification takes place at high temperatures (250°C) with 95 - 97 % yield. Since benzene sulfonic acid (BSA) and H2SO, Card 3/8

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(Table 4) dissolve in the monomer, the latter was heated to 80° 0 in a three-necked flask, and 1 - 5% catalyst was added under vigorous stirring. The resin obtained in a yield of 98% was brittle at room temperature. U.25 M resin obtained in a yield of 98% was brittle at room temperature. U.25 M (49.5 g) monomer in 100 ml toluene yielded with 5% BSA (referred to the monomer) a viscous, rubber-like mass which gradually hardened to a non-melting, unsoluble polymer. All resins were black, with a shining surface, and a specific gravity of 1.1 - 1.5. The dropping point of the resin obtained without BSA was 71°C; that of resin prepared with 1% BSA was 70°C, obtained without BSA was 71°C; that of resin prepared with 1% BSA was 70°C. The resins were found to be well soluble in benzene, its derivatives, drown. The resins were found to be well soluble in benzene, its derivatives, drown. The resins were found to be well soluble in benzene, its derivatives, drown. The resins were found to be well soluble in benzene, its derivatives of toward chlorohydrocarbon, various ketones (cyclohexanone), scarcely in alcohols and ethers. Fractionating allowed recognizing a polydisperse character and ethers. Fractionating allowed recognizing a polydisperse character. Four fractions were separated from a 10% acetone solution: (i) insoluble residue, (2) and (3) were separated by addition of 10 ml H₂0 to a 100 ml

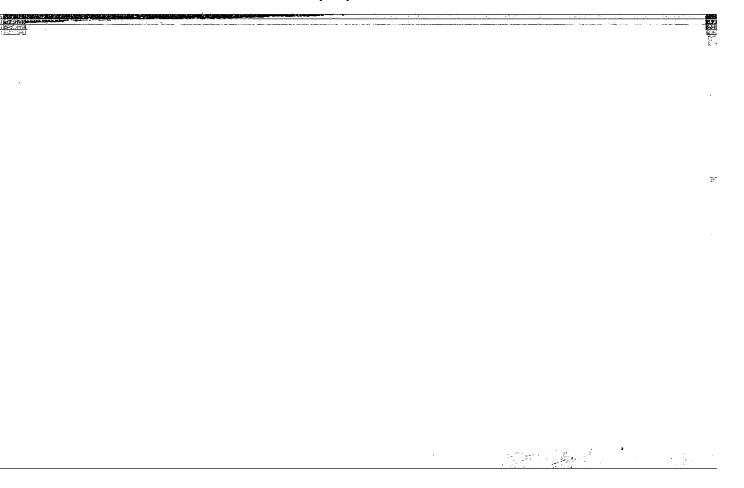
solution, (4) by means of 1000 ml H₂0. Infrared spectra for resing produced without (I) and with (II) catalyst yielded CO hands (1685 1665 cm⁻¹) and double bond bands (1647 - 1621 cm⁻¹) in the conjugate -C=C-O-system. The double bond peaks were, however, found to be weaker particularly with (I). The peak of ethylene bond(1265 - 1310 cm⁻¹) exists only with monomer and (II). The absorption band of the fural ring Card 4/8

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(1131 - 1189 cm⁻¹) is weaker with (I) and (II) than with the monomer. The peaks of the benzene nucleus (1110 - 1070 cm⁻¹) appear in the three spectra, whereas the furan ring bound in a at position (1378 cm⁻¹) was found only, with (I) and (II). There are 1 figure, 5 tables, and 15 references: 6 Soviet-bloc and 9 non-Soviet-bloc. The references to English-language publications read as follows: Ref. 7: US Patent 2,461,508 (1949); Ref. 8: US Patent 2,461,510 (1949); Ref. 9: US Patent 2,768,408 (1956)



GODOVIKOV, N.M.; TERLOV, N.Ye.; KABACHNIK, M.I.

Syntheric of C-ethyl-S-(P-aryloxyethyl) othylthorphogh nates.

Inv.AN SCOR. Ser.khim. nd.1:164-166 166. (MIMA 19:1)

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Role of human ascarids in the epizootiology of ascariasis in piglets. Trudy VIGIS 11:156-160 164. (MIRA 18:12)

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ABUSITOV, S.K., izobretatel; TEPLOV, P.V., izobretatel; GOGULIN, I.Ya., izobretatel;

Designing new looms. Izobr.v SSSR 2 no.2:5-6 F '57. (MIRA 12:3)

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TEPLOV, S.I., kandidat meditsinskikh nauk (Leningrad); SCKOLOVA, Ye.A. (Teningrad)

Effects of the cerebral cortex on the cardiovascular system connected with imminent surgery. Klin.med. 34 no.9:41-47 S 156. (MIRA 9:11)

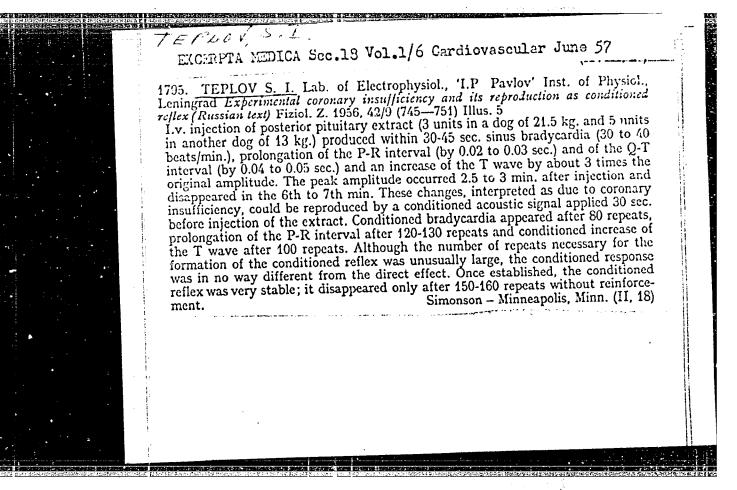
1. Iz terapevticheskogo sektora (zav. deystvitel'nyy chelen AMN SSSR prof. M.V.Chernorutskiy) Instituta fiziologii im. I.P.Pavlova AM 6SSR (dir. akad. K.M.Bykov) i Gospiral'noy khirurgicheskoy kliniki (dir. prof. F.G.Uglov) I Leningradskogo meditsinskogo instituta imeni I.P. Pavlova (dir. A.I.Ivanov)

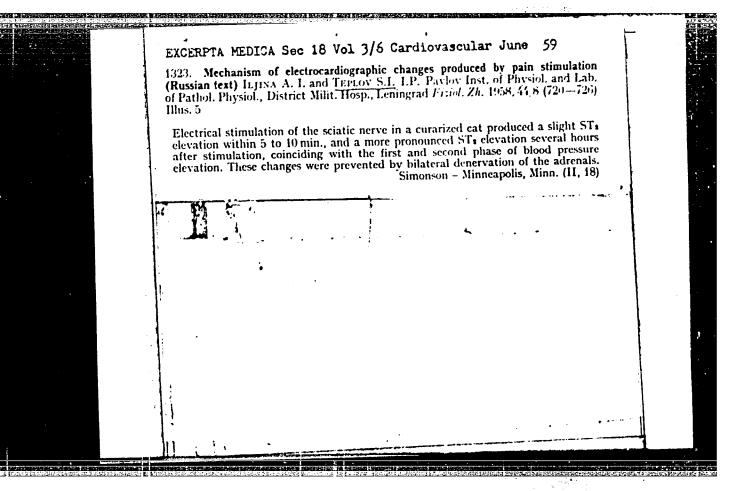
(SURGERY, OPERATIVE, psychol. eff. of cerebral cortex activity on cardiovasc. system)

(CEREBRAL CORTEX, physicl.
eff. of cortical activity on cardiovasc. system before

imminent surg.)
(CARDIOVASCULAR SYSTEM, physical.
eff. of cortical activity befor imminent surg.

APPROVED FOR RELEASE: 07/16/2001 CIA-RDP86-00513R001755310019-9"





Reflex reactions from the stomach on the blood in experimental
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TROLIKIS, A.V., TEPLOV, S.I.

Changes in the secretomotor activity of the stomach in experimental gastritis [with summary in English]. Biul.eksp.biol. i med. 46 (MIRA 11:10) no.8144-48 Ag '58

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eff. on secretomotor activity of stomach in dogs (Rus)) (STOMACH, physiol.

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Mechanisms underlying changes in coronary blood flow accompanying pain stimulation. Fiziol, zhur. SSSR 45 no.7:753-760 Jl '59.

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(CORONARY VESSELS physiology)

(PAIN physiology)

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Pharmacological analysis of the mechanism of changes in the blood pressure and coronary circulation following painful stimulations. Fiziol. zhur. 46 no.12:1456-1462 D '60. (MIRA 14:1)

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Changes in the coronary circulation and blood pressure during stimulation of the hypothalamus region. Fiziol. zhur. 47 no.7: (MINA 15:1) 801-305 Jl 161.

1. From the Laboratory of Tropic Innervation, I.P.Pavlov Institute of Physiology, Leningrad. (BLOOD PRESSURE) (CORONARY VESSELS) (HYPOTHALAMUS)

TEPLOW, Serger Ivanovich; VASILEVSKIY, N.N., red.; SAFRONOVA, I.M., tekhn. red.; KHARASH, G.A., tekhn. red.

[Neural and hormonal regulation of coronary blood circulation]
Nervnaia i gormonal naia reguliatsiia koronarnogo kroboobrashcheNia. Leningrad, Medgiz, 1962. 142 p. (MIRA 15:6)
(CORONARY VESSELS)

L 29213-66 SOURCE CODE: UR/0239/65/051/005/0554/0563 ACC NR: AF6019078 AUTHOR: Toplov. S. I. ORG: Laboratory of the Physiology of the Vegetative Mervous System and Merve Trophics. Institute of Physiology im. I. P. Pavlov, AN SSSR, Leningrad (Laboratoriya fiziologii vegetativnoy nervnoy sistemy i rervnoy trofiki Instituta fiziologii AN SSER) TITIE: Role of adrenergic mechanisms in the development of prolonged changes in the electrocardiogram and blood pressure following stimulation of the hypothalams SOURCE: Fiziologicheskiy zhurnal SSSR, v. 51, no. 5, 1965, 554 - 563 TOPIC TAGS: EKG, blood pressure, cat, vasopressin, hormone ABSTRACT: In experiments conducted on cate, stimulation of the anterior regions of the hypothalamus produced a two-phase (depressor-pressor) reaction of the blood pressure followed by development of a prolonged (up to 3 hours) wave of blood pressure increase. Furthermore, pronounced and stable changes in the EKC, specifically in the ST segment and T wave, were observed. Upon an intravenous injection of chlorpromazine (largactyl) or denervation of the suprarenals, the prolonged pressor reaction and the pressor phase of the initial reaction were absent, while no changes in the EKG occurred. Irritation of the posterior-median hypothalamus generally produced either no change or a gradual decrease in the blood pressure, while the changes in the EKG were minor. In three experiments out of 12, a procipitate drop in the blood pressure took place. The effects of the administration of chlorpromazine or denervation of the suprarchals indicated that the prolonged pressor reaction produced by irritation of the anterior hypothalamus was due to a hormonal reaction initiated by adrenalin and resulting in the evolution of vasopressin by the anterior hypophysis. Blocking of the supply of adrenalin eliminated the pressor reaction. Orig. art. has: 6 figures. [JFRS] SUB CODE: 06/SUBM DATE: 27 Janol ORIGINEF, 005 / OTH REF. 021

SOURCE CCDE: UR/0239/65/051/006/0755/0761 L 28045-66 ACC NR: 1,P6018179 3/ AUTHOR: Tonlikh, A. V.; Il'ina, A. I.; Teplov, S. I. B ORG: Laboratory of Physiology of the Vegetation Nervous System and Norve Trophism, Institute of Physiology im. I. P. Pavlov, AN SSSR, Leningrad (Laboratoriya fiziologii vegetativncy nervnoy sistemy i nervnoy trofiki Instituta fiziologii AN SSSR) TITLE: Changes in the electrical activity of the hypothalamus upon irritation of a sensory nerve or administration of adrenaline SOURCE: Fiziologicheskiy zhurnal, v. 51, no. 6, 1965, 755-761 TOPIC TAGS: pharmacology, electrophysiology, cat, EEG, brain, blood pressure, rasopressin, animal physiology ABSTRACT: In experiments on cats, irritation of the central end of a severed sciatic nerve (a pain irritation) and intravenous injection of adrenaline had the same effect on the electric activity of the hypothalamus: the activity in both the anterior and posterior divisions of the hypothalamus was increased (desynchronization of EEG rhythms took place and the amplitude of EEG waves was increased). This reaction coincided with an increase in the blood pressure, but was senetimes accompanied by a blood pressure decrease. Within 1.5-3 hrs. after the primary effect (stimulation of the electrical activity of the hypothalamus following the pain UDC: 612.822.3.087 Card 1/2

L 28045-66 ACC NR. AP6018179 [irritation or injection of adrenaline), a second increase in the electrical activity of the hypothalanus took place, which coincided with the prolonged wave of blood pressure increase described in the authors' earlier work. One may assume that a chain neurohormonal reaction involving stimulation of the hypothalamus doveloped both in response to irritation of the sciatic nerve and to injection of adrenaline. Irritation of the sciatic nerve stimulated the sympathico-adrenal system; vasoconstriction under the effect of nerve action and also release into the blood of adrenaline and vasopressin, which was controlled by the vegetative centers of the hypothalamus, took place. The initially released adrenaline stimulated the hypothalamus, with the result that vasopressin was released, producing the second, prolonged increase in blood pressure, which was of purely normonal origin. Orig. SUB CODE: 06/ SUBM DATE: 30Jan64/ ORIG REF: 005/ OTH REF: 009 2/2 CC

VASILIYEVA, L.I.; TEPLOY, S.I.

Changes in the coronary blood flow in stimulation of the afferent fibers of the vagus nerve. Fiziol.zhur. 51 no.7:326-331 (MIRA 18:10)

1. Iaboratoriya fiziologii vegetativnoy nervnoy sistemy i nervnoy trofiki Instituta fiziologii imeni 1.P.Pavlova AN SSSR, Leningrad.

YELIZAROV, P.P., kandidat tekhnicheskikh nauk; TEPLOV, S.V., inzhener
Heat losses during the starting and shutdown of the TP-170
boiler. Teploenergetika 2 no.7:38-44 J1 55. (MIRA 8:10)

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BLOKHIN, V.N.; GRIGOR'YEV, M.G.; KOZHEVNIKOV, A.I.; KOROLEV, B.A.; MATYUSHIN, I.F.: PARIN, B.V.; TSIMKHES, I.L.; KALINIMA, G.V.; FEDOROV, A.M.; KOLOKOL'TSEV, M.V.; SOKOLOV, V.V.; PRILUCHMAYA, O.A.; SHUMILKINA, YE.I.; ABRAMOV, Yu.G.; HIURIKOV, A.Kh.; IKONNIKOV, P.I.; VOZNESENSKIY, I.Ya.; TEPLOV. S.V.; MIZINOV, N.N.; KUKOSH, V.I.

V.M.Durmashkin; obituary. Ortop., travm. 1 protez. 21 no.8:81 Ag (MIRA 13:11)

'60. (DURMASHKIN, VIKTOR MARKOVICH, d. 1960)

ACCESSION NR: AP4004156

\$/0294/63/001/002/0318/0320

AUTHOR: Filimonov, S. S.; Kryukova, M. G.; Teplov, S. V.; Ayristov, A. A.

TITLE: Test stand for studying heat transfer in the flow of liquid aluminum in a pipe

SOURCE: Teplofizika vy*sokikh temperatur, v. 1, no. 2, 1963, 318-320

TOPIC TAGS: heat transfer, liquid aluminum heat exchanger, liquid metal, liquid aluminum, aluminum heat transfer, heat exchanger, liquid metal coolant, coolant, fluid flow

ABSTRACT: A test stand has been designed for heat-transfer studies with liquid aluminum. The use of liquid aluminum as a heat-transfer agent in heat exchangers operating at temperatures exceeding 1200C is being investigated since difficulties are encountered with alkali metals at such temperatures. Fig. 1 of the Enclosure shows the test assembly. An induction-type electromagnetic pump with a traveling magnetic field (capacity 3 m³/hr) was specially Cord 1/3-